

**AMENDMENT TO THE SPECIFICATION**

Please amend the specification as follows:

On page 1, please amend the title as follows:

**INVERTER ~~DRIVING~~ DRIVER ~~DEVICE~~ AND METHOD**

[0008] However, in the inverter driver according to the conventional duty control method, the current waveform of the CCFL is steeply varied when the duty greatly reduces or increases, ~~and~~. Accordingly, the brightness of the CCFL becomes unstable, interference occurs in the adjacent circuit because of many harmonics thereof, and the lifetime of the CCFL shortens.

[0021] FIG. 1 shows an inverter driver according to a first preferred embodiment of the present invention;

[0022] FIG. 2 shows a ratio of a voltage V1 versus a voltage V2, and a relation between a voltage Vct and a frequency f in the inverter circuit 100;

[0023] FIG. 3 shows a signal waveform diagram according to a first preferred embodiment of the present invention; ~~and~~.

[0029] The inverter circuit 100 uses a serial/parallel resonance of a half bridge inverter; ~~and a.~~ A resonance frequency of the inverter circuit 100 is a frequency whereby the total impedance of the inductor L1 and the capacitors C1, C2, and C3 becomes zero in the viewpoint of from the primary side to the secondary side of the inverter circuit 100.

[0030] Body diodes D1 and D2 are respectively coupled to the switches M1 and M2 of the inverter circuit 100, ~~and the.~~ The body diodes enable zero voltage switching of the switches M1 and M2 as described later.

[0031] The control signal supply 200 comprises resistors R1 and R2 coupled in series

between the input voltage  $V_{cc}$  and ground; ~~a.~~ A subtractor 220 ~~for subtracting~~ a voltage  $V_{nc}$  at a node between the resistors  $R1$  and  $R2$  from a reference voltage  $V_r$  and outputting a subtraction voltage  $V_a$  ( $V_a = V_r - V_{nc}$ ); ~~a.~~ A comparator 240 ~~for comparing~~ a reference voltage  $V_{ref}$  and a feedback voltage  $V_{fb}$  at a resistor  $R_{sense}$  sensing the current flowing to the CCFL 10, ~~amplifying~~ the comparison result, and ~~outputting~~ a voltage  $V_{comp}$ ; ~~and a.~~ A multiplier 260 ~~for multiplying~~ output signals of the subtractor 220 and the comparator 240 by a predetermined gain  $K$  to generate a voltage  $V_{mo}$ , and ~~supplying~~ the voltage  $V_{mo}$  to the duty controller 300.

[0034] The comparator 310 compares the output voltage  $V_{mo}$  of the control signal supply 200 with a voltage  $V_{ct}$  charged in the capacitor  $C_t$  of the frequency controller 400, and provides a comparison result to the R ~~end~~ terminal of the RS latch 320. The S ~~end~~ terminal of the RS latch 320 receives clock signals CLK from an oscillator 410 of the frequency controller 400. Signals output from the Q' ~~end~~ terminal of the RS latch 320 and the clock signals CLK of the oscillator 410 are input to two input ~~ends~~ terminals of the OR/NOR logic gate 330. Two output signals of the OR/NOR logic gate 330 are respectively provided to the high-side gate driver 350 for driving the switch  $M1$  and the low-side gate driver 340 for driving the switch  $M2$ .

[0038] When the voltage  $V_t$  obtained by subtracting the voltage  $V_x$  from the voltage  $V_{comp}$  is greater than the voltage  $V_{rt}$ , the current  $I_{C1}$  flows to the resistor  $R_f$ , and the current  $I_{Ct}$  which is the difference  ~~$I_{C2} - I_{C1}$~~  between the currents  $I_{C1}$  and  $I_{C2}$  (i.e.,  $I_{C2} - I_{C1}$ ) flows to a terminal of the oscillator 410 to which the resistor  $R_t$  is coupled.

[0039] The capacitor  $C_t$  is coupled to the oscillator 410, ~~and since.~~ Since the current flowing to the capacitor  $C_t$  is matched with the current  $I_{Ct}$ , the current  $I_{Ct}$  charges or discharges the voltage at the capacitor  $C_t$ .

[0041] Given an amplitude  $V$  of the voltage  $V_{ct}$ , the period of the voltage  $V_{ct}$  charged in the capacitor  $C_t$  is the summation of the charge time  $(C_t \times V)/I_{Ct}$  and the discharge time  $(C_t \times V)/I_{Ct}$ , ~~and accordingly.~~ Accordingly, the frequency  $f$  of the voltage  $V_{ct}$  is given as

Equation 2.

[0046] The operation frequency region of the inverter driver is between the minimum frequency  $f_{low}$  and the maximum frequency  $f_{high}$ , ~~and as~~. As given in Equation 2, since the capacitor  $C_t$  is constant and the amplitude  $V$  of the voltage  $V_{ct}$  is also constant, the maximum frequency  $f_{high}$  is obtained when the current  $I_{ct}$  is a maximum, and the minimum frequency  $f_{low}$  is obtained when the current  $I_{ct}$  is a minimum.

[0047] Since  $I_{ct} = I_{c2} - I_{c1}$  and  $I_{c2} = V_{rt}/R_t$ , the  $I_{ct}$  becomes the maximum and the frequency of the voltage  $V_{ct}$  accordingly becomes the maximum frequency  $f_{high}$  when  $I_{c1} = 0$ , ~~and~~.  $I_{ct}$  becomes the minimum and the frequency of the voltage  $V_{dt}$  becomes the minimum frequency  $f_{low}$  when  $I_{c1}$  is the maximum.

[0069] The output voltage  $V_{comp}$  of the comparator 240 is input to ~~a non-an~~ an inverting end of the OP amp 430, and a resistor  $R_f$  is coupled between an inverting end of the OP amp 430 and the ground voltage. Since the voltages at the inverting and non-inverting ends of the OP amp 430 are the same, the voltage  $V_{comp}$  is applied to both ends of the resistor  $R_f$ , and the current  $I_{c1}$  flowing to the resistor  $R_f$  is  $V_{comp}/R_f$ .